

METHOD AND APPARATUS FOR OPTICAL TRANSMISSION

CROSS REFERENCE TO RELATED APPLICATIONS

This application claims priority of Provisional Application Serial No.
5 60/546504 which was filed February 20, 2004.

FIELD OF THE INVENTION

The present invention generally relates to the field of optical communications, and more particularly to optical communication apparatus for generating alternate-polarization phase-shift-keyed optical signals.

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BACKGROUND OF THE INVENTION

In highly dispersed pseudo-linear high bit-rate long-haul transmission systems, intra-channel four-wave mixing (IFWM) and intra-channel cross-phase modulation (IXPM) are the main causes for nonlinear penalties. Using a
15 Differential-phase-shift-keying (DPSK) modulation format can mitigate these penalties as compared to on-off-keying (OOK). Several 40-Gb/s DPSK ultra-long-haul DWDM transmission experiments have shown excellent capacity and distance performance. The strong resistance of DPSK against IXPM stems from the equal pulse energy in each time slot. It has also been shown that the reduced
20 pulse energy of Return-to-Zero (RZ)-DPSK compared to RZ-OOK and a correlation between the nonlinear phase shifts of two adjacent pulses contribute to the better robustness of RZ-DPSK against IFWM.

In principle, alternating the polarization of the signal from bit to bit will further reduce the nonlinear penalties since there is little FWM between two orthogonally
25 polarized signals and the effect of XPM between the two orthogonal polarizations is half that between the parallel ones. Prior art simulations have shown that more than 50% improvement in the transmission distance can be achieved by using

Alternate Polarization-DPSK (APol-DPSK) compared to single polarization DPSK. The simultaneous application of DPSK and polarization alternation enhances the performance in an optical transport system where the intra-channel optical nonlinear interactions are limiting factors.

5 One prior art method for implementing alternate-polarization DPSK uses optical-time-division-multiplexing (OTDM) and polarization-multiplexing in the transmitter. The method requires a short-pulse source operating at half bit rate, OTDM demultiplexing or polarization demultiplexing at receiver, and complicated transmitters and receivers. Such a method is not practical in real system
10 applications and is not suitable for Non-Return-to-Zero (NRZ) formats. This approach is also impractical for optical networks because it requires multiple optical transmitters for a single channel.

Moreover, prior art methods of DPSK data precoding and encoding are not suitable for efficiently generating alternate polarization formatted signals.

15 **SUMMARY OF THE INVENTION**

These and other deficiencies of the prior art are addressed by embodiments of the present invention that include methods and apparatus for generating and/or transmitting Alternate Polarization-DPSK (APol-DPSK) optical signals, where information is optically encoded between optical bits of an optical signal temporally
20 separated by an even number of bit periods. The present invention can be applied to DPSK and PSK formats, including but not limited to, Differential-Quadrature-Phase-Shift-Keying (DQPSK), Differential- $\pi/2$ -Phase-Shift-Keying (D- $\pi/2$ -PSK), and the like.

In one embodiment, a method of APol-PSK transmission is provided which
25 includes modulating the output of an optical source to optically encode electronic data using phase shift keying (PSK) to generate an optical signal, and alternating

the polarization of the phase shift keyed optical signal using a modulator such that successive optical bits have substantially orthogonal polarizations to generate an APol-PSK signal.

In another embodiment, a method of APol-DPSK transmission is provided
5 which includes precoding an electronic data signal, modulating the output of a laser using the precoded electronic data signal, and differential phase shift keying between two optical bits separated by an even number of bit periods to generate an optical signal. The polarization of the differentially phase shift keyed optical signal is alternated using a modulator such that successive optical bits have
10 substantially orthogonal polarizations to generate an APol-DPSK signal.

In another embodiment an optical transmitter for APol-DPSK transmission is provided comprising an optical source, and a precoder device for precoding an electronic data signal. An optical phase-shift-keying data modulator is driven by a precoded electronic data signal from the precoder device to produce an optical
15 DPSK signal such that the signal is carried as phase shift keying between two optical bits separated by an even number of bit periods. A polarization alternator is optically coupled to the data modulator to provide polarization alternation of the output of the data modulator. .

In another embodiment, an optical transmitter for APol-DPSK transmission
20 is provided comprising an optical source, a precoder device for precoding an electronic data signal, and a modified Mach-Zehnder (MZ) modulator having a polarization rotation device in at least one arm. Drive circuitry for the MZ modulator is provided to process the precoder output and to simultaneously achieve polarization alternation and optical data encoding as phase shift keying
25 between the two optical bits separated by an even number of bit periods.

BRIEF DESCRIPTION OF THE DRAWINGS

The foregoing and other objects, features and advantages of the present invention will become more apparent from the following detailed description when taken in conjunction with the appended drawings.

5 It is to be noted, however, that the appended drawings illustrate only exemplary embodiments of the invention and are therefore not to be considered limiting of the scope of the invention.

FIG. 1 is an illustration of an RZ-APol-DPSK formatted data stream encoded with '101101';

10 FIG. 2A is a diagram showing a 40-Gb/s APol-DPSK precoder that can be used in accordance with embodiments of the invention;

FIG. 2B is a diagram showing a 40-Gb/s APol-DPSK precoder integrated with a 10-Gb/s to 40-Gb/s electronic time-division multiplexer that can be used in accordance with embodiments of the invention;

15 FIG. 3A shows an embodiment of an RZ-APol-DPSK transmission apparatus using a DFB: distributed feedback laser, PC: RZ pulse carver, PM: phase modulator, PA: polarization alternator in accordance with aspects of the invention;

FIG. 3B shows an embodiment of an APol-DPSK transmission apparatus
20 using a single modulator for simultaneous data modulation and polarization alternation;

FIGs. 4 A-C are drawings illustrating alternate embodiments of polarization alternating modulators that may be used in accordance with the present invention;

FIG 5 is a plot which illustrates orientations of the electric field of light output
25 of the device shown in FIG 4 B as a function of the drive voltages; and

FIG. 6 is a logic table showing the relations among an APol-DPSK precoded bit stream D for the data '101101', a 20G Clock (CLK), and a phase section drive voltages for the device in FIG. 4 B.

DETAILED DESCRIPTION

Although the present invention is discussed below with reference to two-bit delay implementations of APol-DPSK, it can be understood by those skilled in art that straightforward modifications allow implementing APol-DPSK for other even-
5 bit delay cases. It is also understood that the present invention can be applied to accomplish polarization alternation with other phase-shift-keying formats, including but not limited to DQPSK, PSK (non-differential) and D- $\pi/2$ -PSK.

APol-DPSK data encoding: In conventional optical DPSK transmission, information is encoded by 0- or π -phase shifting between successive bits. In
10 accordance with the present invention this strategy is modified so alternate polarization may be used with DPSK (since successive bits have orthogonal polarizations, they do not interfere in a conventional DPSK demodulator that uses a one-bit delay line interferometer). The encoding for APol-DPSK may be accomplished by differentially phase shift keying between any two bits separated
15 by an even number of bit slots (having the same polarization). Two-bit delay is the shortest possible and is preferred owing to the consequent simplicity of the implementation and to the increased polarization and phase correlation between the two bits. The phase and polarization information of two-bit delay APol-DPSK is illustrated in FIG. 1. The demodulation of the signal can then be achieved using,
20 for example, an even bit delay-line interferometer.

APol-DPSK precoding: A precoder, such as those shown in FIGs. 2 A-B, is preferably used in systems for generating APol-DPSK optical signals according to the present invention. The role of the precoder is to process the input data in such a manner that the output of the precoder may be optically encoded as the phase-
25 shift-keying between two optical bits separated by an even number of bit periods.

The precoders shown in FIGs. 2A-B utilize the fact that a stream of APol-DPSK formatted data can be logically split into two independent sets of orthogonal

polarizations, where each set is, for example, a conventional 20 Gb/s DPSK. It is to be noted that the AND and Toggle Flip-Flop (T-FF) elements in FIGs. 2 A-B need only to operate at half the bit rate (e.g. 20 Gb/s), in contrast to the comparable elements in a conventional (e.g. 40-Gb/s) DPSK precoder.

5 In practice, a 40-Gb/s data stream can be electronically multiplexed out of lower bit-rate tributaries, and the precoder can be integrated in the multiplexing process, as depicted in Fig. 2B for the case of electronic time-division multiplexing (ETDM) of four 10-Gb/s data streams.

APol-DPSK optical modulation: The optical modulation and transmission of
10 the precoded electronic data for APol-DPSK according to the invention can be achieved by, for example, the implementations shown in FIGs. 3A-B. Figure 3 A depicts an optical transmission system 300 according to one embodiment of the invention where a polarization alternator (PA) 310 is used for alternating the polarizations of a modulated optical signal from a phase-shift-keying data
15 modulator (PM) 320. The phase-shift-keying data modulator can be, for example, either a phase modulator, or a Mach-Zehnder modulator.

Several designs of PAs 310 are possible. One example of such a PA is a phase modulator (shown in FIG. 4A). The operational principle of the PA of FIG. 4A is similar to that of an electrooptic polarization scrambler. When an optical
20 signal is launched into the modulator with its polarization oriented at 45 degrees with respect to the crystal axis, the output polarization is $\hat{x} \exp(jrV(t)) + \hat{y}$, where r is the difference between the TE and TM electrooptic coefficients and $V(t)$ is the voltage applied to the phase modulator. The drive voltage is preferably a sine wave or square wave with the repetition rate half of the bit rate and the peak-to-
25 peak amplitude of the voltage should produce π phase difference between the TE and TM polarization components, thereby producing polarization alternation from bit to bit. Note that the data modulator and polarization alternator can be easily

integrated into a single device as noted by the dotted box in FIG. 3A. In this case, an integrated wave plate can be inserted between the two modulator sections in order to rotate the polarization of the output of the data modulator section by 45 degrees. If there exists a substantial difference between the propagation velocities of TE and TM modes in the polarization alternator, a birefringent element may be inserted to compensate for the difference.

FIG 4B illustrates a modified Mach-Zehnder modulator device having a polarization rotation component, for example a half-wave plate, integrated in one arm to provide rotation of the polarization of the light in the arm by 90 degrees. The light outputs from the two arms are combined by a polarization combining or 3-dB power combining coupler junction. Such a device can be driven in a similar manner to the device in FIG. 4A to achieve polarization alternation. The difference of the drive voltages applied to the two arms of the modulator is preferably a sine wave or square wave with the repetition rate half of the bit rate, and the amplitude of the difference voltage is set to V_π of the modulator.

Fig. 4C illustrates a Mach-Zehnder modulator device having two complementary output ports. A polarization beam combiner (PBC) is also provided for combining signals from the two output ports. Such a device can be driven in a similar manner to the device in Fig. 4B to achieve polarization alternation.

FIG 3B shows an optical transmission system 350 according to another embodiment of the invention in which the polarization alternation and the data modulation are accomplished in a single optical modulator 360, as shown in FIG 4B. A modified Mach-Zehnder modulator is preferably used, where the polarization of the field in one arm is rotated by 90 degrees. Assuming that the input field is polarized along \hat{x} , the field in the upper arm of the interferometer in FIG 4B is $\hat{x} \exp(j\pi V_1(t)/V_\pi)$ while the field in the lower arm is $\hat{y} \exp(j\pi V_2(t)/V_\pi)$ after the polarization rotation. Those skilled in the art will appreciate that these fields can be recombined by, for example, a polarization combining coupler

junction or a 3-dB power combining coupler junction.

It will be apparent to those skilled in the art that other ways of providing the polarization rotation in one arm of the interferometer as well as other way of recombining the two arms would lead to substantially identical operation.

- 5 Additionally, it should be apparent that other ways of modifying the state of polarization of the two arms of the interferometer would be suitable as long as they provide substantially orthogonal states of polarization before recombination

The orientations of the electric field vector for the output of the modulator 360 can be written as $\hat{x}\exp(j\pi V_1/V_\pi) + \hat{y}\exp(j\pi V_2/V_\pi)$, and are illustrated in Fig. 5
 10 for four combinations of the voltages applied to the two electrodes of the modulator 360. It can be understood from FIG. 1 and FIG. 5 that the sequence of the pairs of voltages V_1 and V_2 in the table of FIG. 6 yields the APol-DPSK encoded bit stream for '101101' shown in FIG 1. Moreover, the voltages V_1 and V_2 are simply related to the precoded data (D) and 20 G clock, namely $V_1/V_\pi = D \oplus \text{CLK}$, and $V_2/V_\pi = D$,
 15 where \oplus denotes XOR operation.

In order to demonstrate aspects of the present invention and the performance improvement of APol-DPSK over DPSK, a single-channel 42.7-Gb/s transmission experiment was performed using an implementation similar to that shown in FIG.3A. A 1551-nm laser and pulse carver produced 8-ps optical pulses,
 20 which were then DPSK-modulated using a Mach-Zehnder modulator. Data precoding was not necessary, since a pseudo-random data sequence (length $2^{31} - 1$, electronically multiplexed from four copies of a 10.7-Gb/s sequence with quarter-pattern delays) was used. The 330-km fiber link consisted of four ~ 82-km spans of standard single-mode fiber (SSMF), each followed by dispersion-
 25 compensating fiber (DCF). A Raman +Erbium hybrid amplification scheme was used in each span. The launched power in each span was +7 dBm. The received optical signal-to-noise ratio was set to 22.2 dB (measured in a 1-nm resolution

bandwidth and referred to 0.1 nm). A 4-bit (93.7-ps) delay interferometer was used as the demodulator. This was followed by single-photodiode detection, clock recovery, and electronic demultiplexing to 10.7 Gb/s for bit-error-rate (BER) testing. Without alternating the polarization, the BER was 10^{-6} . A sinusoidally
5 driven polarization alternator was then added to the transmitter, and the BER improved to 10^{-9} .

An additional demonstration of aspects of the invention was performed and is discussed in A.H. Gnauck et al., Paper PDP 35, "6 × 42.7-Gb/s transmission over ten 200-km EDFA-amplified SSMF spans using polarization-alternating RZ-
10 DPSK", OFC 2004, February 22-27 2004, which is incorporated herein.

Although the invention has been described with reference to illustrative embodiments, this description should not be construed in a limiting sense. Various modifications of the described embodiments, as well as other embodiments of the invention, which are apparent to persons skilled in the art to which the invention
15 pertains, are deemed to lie within the principle and scope of the invention.

Although the steps of various methods according to the invention are disclosed and discussed with reference to a particular sequence, the present invention is not limited to methods implementing some or all of those steps in those particular sequences.